

Sub-THz Band Transmission Technology for 6G Wireless Communication

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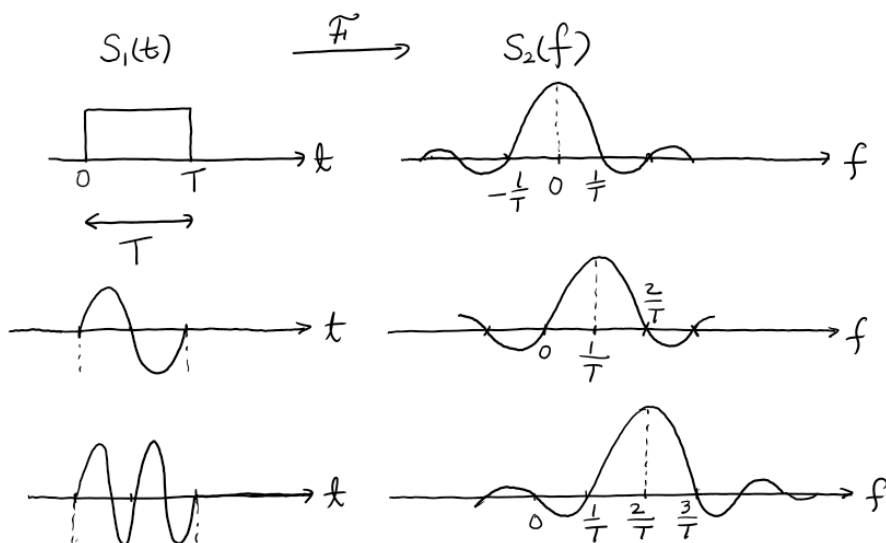
- 3G wireless communication
 - ✓ Time resource: 2 MHz bandwidth at 1 GHz carrier frequency
 - ✓ Spectral efficiency : 1 bps/Hz, Capacity: 2 Mbps
- 4G wireless communication
 - ✓ Time resource: 20 MHz bandwidth at 2 GHz carrier frequency
 - ✓ Space resource (MIMO): number of antennas = 4
 - ✓ Spectral efficiency : 15 bps/Hz, Capacity: 300 Mbps
- 5G wireless communication
 - ✓ Time resource: 1 GHz bandwidth at 28 GHz carrier frequency
 - ✓ Space resource: number of beams * polarization
 - ✓ Spectral efficiency : 20 bps/Hz, Capacity: 20 Gbps

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Introduction

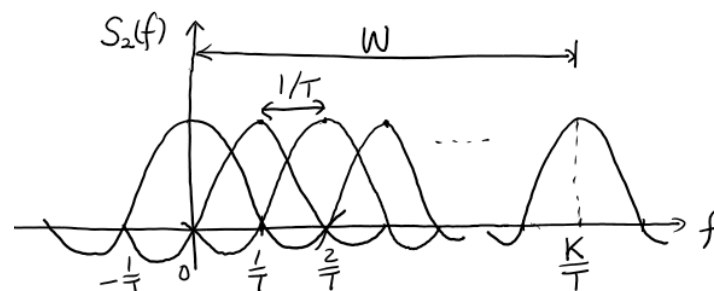
- Number of information from time and frequency resource
 - ✓ Time resource: T , frequency resource: W
 - ✓ Time and frequency have fourier transform relationship.
 - ✓ Number of sinc functions in frequency domain

$$K = \frac{W}{1/T} = WT \text{ (Number of information)}$$



$$S_2(f) = \int_{-\infty}^{\infty} S_1(t) e^{-j2\pi ft} dt$$

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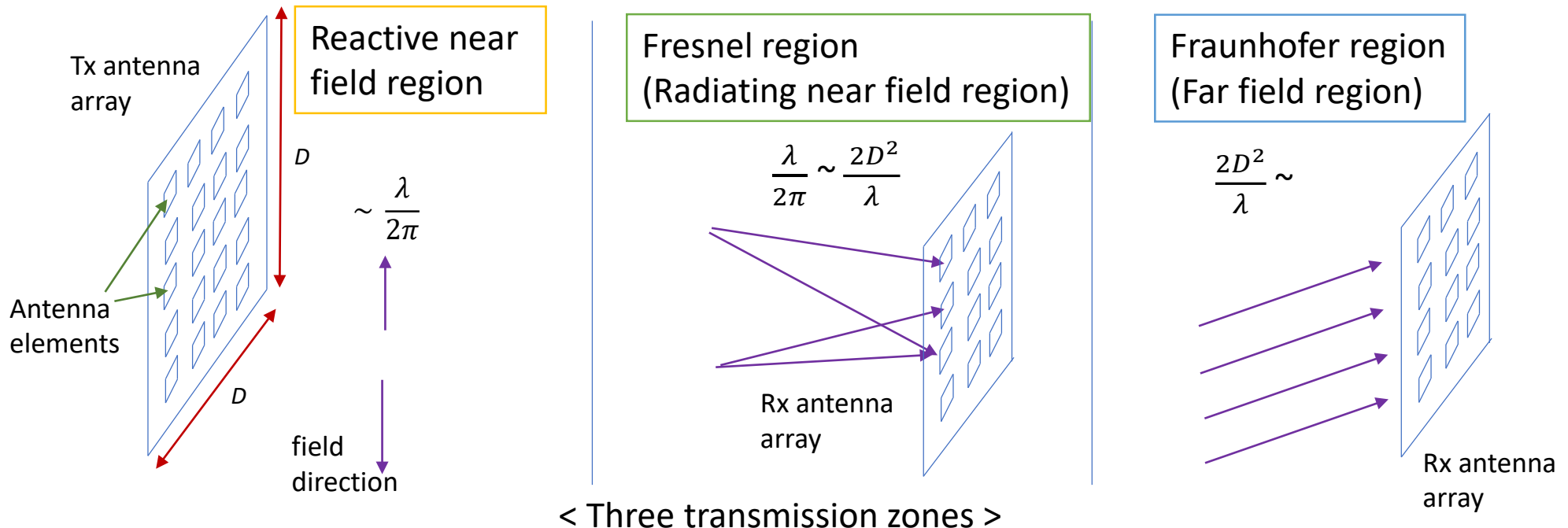
< Utilization of time and frequency resources >

- 6G capacity requirement
 - ✓ The goal of 6G: 1 Tbps capacity
- 6G frequency resource and capacity
 - ✓ Sub-THz band: 100 GHz~300 GHz
 - ✓ Available bandwidth: 5 GHz~10 GHz
 - ✓ Required spectral efficiency: 100~200 bps/Hz
 - ✓ $(64\text{QAM}) * (\text{polarization}) = (6 \text{ bps/Hz}) * 2 = 12 \text{ bps/Hz}$
- Can we achieve the spectral efficiency of 100~200 bps/Hz for 6G 1 Tbps capacity?
 - ✓ Yes, capacity can be increased by utilizing spatial resources in the Fresnel region.

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Fresnel Region Transmission

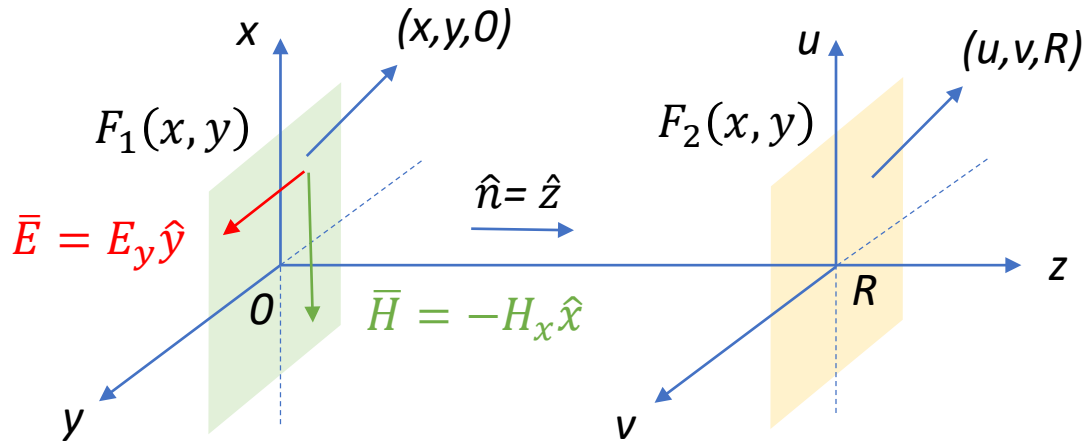
- Transmission region in a LoS (Line-of-sight) environment
 - ✓ Reactive near field region: Energy is stored without radiation.
 - ✓ Fresnel region: Different pairs between Tx antenna elements and Rx antenna elements have different channel characteristics.
 - ✓ Fraunhofer region: The field reaches all receiver antenna elements in the same direction.



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Fresnel Region Transmission

- Relationship between Tx field and Rx field (Fresnel approximation)



$$\bar{J}(\bar{r}) = \hat{n} \times \bar{H} = \hat{z} \times (-H_x \hat{x}) = -H_x \hat{y}$$

$$\bar{M}(\bar{r}) = -\hat{n} \times \bar{E} = -\hat{z} \times (E_y \hat{y}) = E_y \hat{x}$$

$F_1(x, y)$: Tx field (E or H)

$F_2(x, y)$: Rx field (E or H)

$$\bar{r} = (x, y, 0), \bar{r}' = (u, v, R)$$

$$|\bar{r} - \bar{r}'| = \sqrt{(x - u)^2 + (y - v)^2 + R^2}$$

$$F_2(u, v) = \frac{j}{\lambda} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F_1(x, y) \frac{e^{-jk|\bar{r} - \bar{r}'|}}{|\bar{r} - \bar{r}'|} dx dy$$

$$\approx \frac{je^{-jkR}}{\lambda R} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F_1(x, y) \exp\left(-\frac{jk}{2R}(u - x)^2 - \frac{jk}{2R}(v - y)^2\right) dx dy$$

$$= \frac{je^{-jkR}}{\lambda R} e^{-\frac{jk}{2R}(u^2 + v^2)} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F_1(x, y) e^{-\frac{jk}{2R}(x^2 + y^2)} e^{\frac{jk}{R}(ux + vy)} dx dy$$

- Transformation from signals to fields

$S_1(x, y)$: Tx signal, $S_2(u, v)$: Rx signal

$$F_1(x, y) = S_1(x, y) e^{\frac{jk}{2R}(x^2+y^2)}, \quad F_2(u, v) = S_2(u, v) e^{-\frac{jk}{2R}(u^2+v^2)}$$

- Relationship between Tx signal and Rx signal

✓ Tx/Rx signals have a 2-dimensional Fourier transform relationship.

$$\begin{aligned} S_2(u, v) &= \frac{je^{-jkR}}{\lambda R} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S_1(x, y) e^{j\frac{k}{R}(ux+vy)} dx dy \\ &= je^{-jkR} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S_1(x, y) e^{j2\pi \left[\left(\frac{u}{\sqrt{\lambda R}}\right) \left(\frac{x}{\sqrt{\lambda R}}\right) + \left(\frac{v}{\sqrt{\lambda R}}\right) \left(\frac{y}{\sqrt{\lambda R}}\right) \right]} \frac{dx}{\sqrt{\lambda R}} \frac{dy}{\sqrt{\lambda R}} \end{aligned}$$

$$S_1(x, y) = -je^{jkR} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S_2(u, v) e^{-j2\pi \left[\left(\frac{x}{\sqrt{\lambda R}}\right) \left(\frac{u}{\sqrt{\lambda R}}\right) + \left(\frac{y}{\sqrt{\lambda R}}\right) \left(\frac{v}{\sqrt{\lambda R}}\right) \right]} \frac{du}{\sqrt{\lambda R}} \frac{dv}{\sqrt{\lambda R}}$$

$$S_1(x, y) \xleftrightarrow{\text{2-dim. Fourier transform}} S_2(u, v)$$

- Two one-dim. Fourier transform

✓ If $x' = \frac{x}{\sqrt{\lambda R}}$, $y' = \frac{y}{\sqrt{\lambda R}}$, $u' = \frac{u}{\sqrt{\lambda R}}$, $v' = \frac{v}{\sqrt{\lambda R}}$,

then $S_2(u', v') = j e^{-jkR} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} S_1(x', y') e^{j2\pi(u'x' + v'y')} dx' dy'$

✓ If $S_1(x', y') = S_{1,H}(x') S_{1,V}(y')$, $S_2(u', v') = j e^{-jkR} S_{2,H}(u') S_{2,V}(v')$,

then $S_{2,H}(u') = \int_{-\infty}^{\infty} S_{1,H}(x') e^{j2\pi u'x'} dx'$, $S_{2,V}(v') = \int_{-\infty}^{\infty} S_{1,V}(y') e^{j2\pi v'y'} dy'$

- Spatial resources

✓ Tx array size: $A_1 * B_1$, Rx array size: $A_2 * B_2$, distance: R , wavelength: λ

✓ Tx space resource: $0 \leq x' \leq \frac{A_1}{\sqrt{\lambda R}}$, $0 \leq y' \leq \frac{B_1}{\sqrt{\lambda R}}$

✓ Rx space resource: $0 \leq u' \leq \frac{A_2}{\sqrt{\lambda R}}$, $0 \leq v' \leq \frac{B_2}{\sqrt{\lambda R}}$

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Fresnel Region Transmission

- Time and frequency resource → Fourier transform relationship
 - ✓ Time resource T , and frequency resource W
 - ✓ Number of information: $T*W$
- Linear array resource → Fourier transform relationship
 - ✓ Tx array resource $\left(\frac{A_1}{\sqrt{\lambda R}}\right)$, and Rx array resource $\left(\frac{A_2}{\sqrt{\lambda R}}\right)$
 - ✓ Number of modes for linear array: $\left(\frac{A_1}{\sqrt{\lambda R}}\right)\left(\frac{A_2}{\sqrt{\lambda R}}\right) = \frac{A_1 A_2}{\lambda R}$
- Planar array resource → 2-dim. Fourier transform relationship
 - ✓ Number of modes for horizontal(vertical) direction array: $\frac{A_1 A_2}{\lambda R}, \frac{B_1 B_2}{\lambda R}$
 - ✓ Number of modes for planar array: $\left(\frac{A_1}{\sqrt{\lambda R}}\right)\left(\frac{B_1}{\sqrt{\lambda R}}\right)\left(\frac{A_2}{\sqrt{\lambda R}}\right)\left(\frac{B_2}{\sqrt{\lambda R}}\right) = \frac{A_1 A_2 B_1 B_2}{(\lambda R)^2}$

(Example)

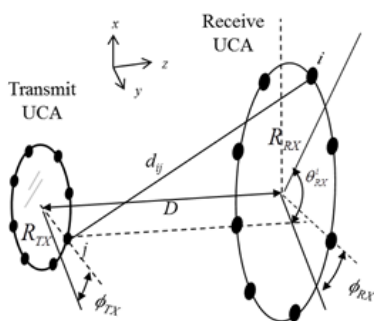
Carrier frequency: 150 GHz , $\lambda = 2 \text{ mm}$, $R = 3 \text{ m}$, $A_1 = A_2 = B_1 = B_2 = 20 \text{ cm}$

Number of modes: $\frac{20\text{cm}*20\text{cm}*20\text{cm}*20\text{cm}}{(3\text{m}*2\text{mm})^2} = 44$

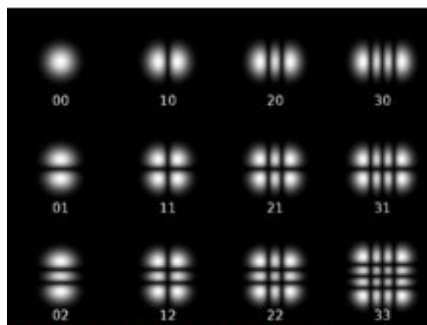
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Fresnel Region Transmission

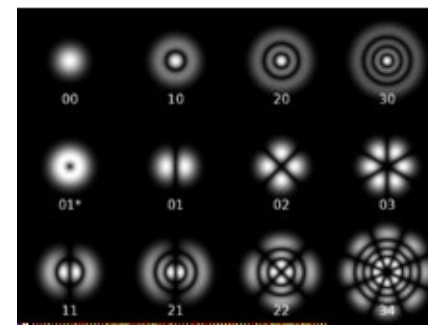
- System design for spatial multimode transmission
 - ✓ Channel environment: LoS environment
 - ✓ In optics, various multi-mode transmission schemes such as OAM, Hermite Gaussian and Laguerre Gaussian modes, have been studied.
 - ✓ In order to apply it to wireless communication environment, it is necessary to develop a transmission scheme that addresses RF impairment problems and supports user mobility.



< OAM mode >



< HG mode >

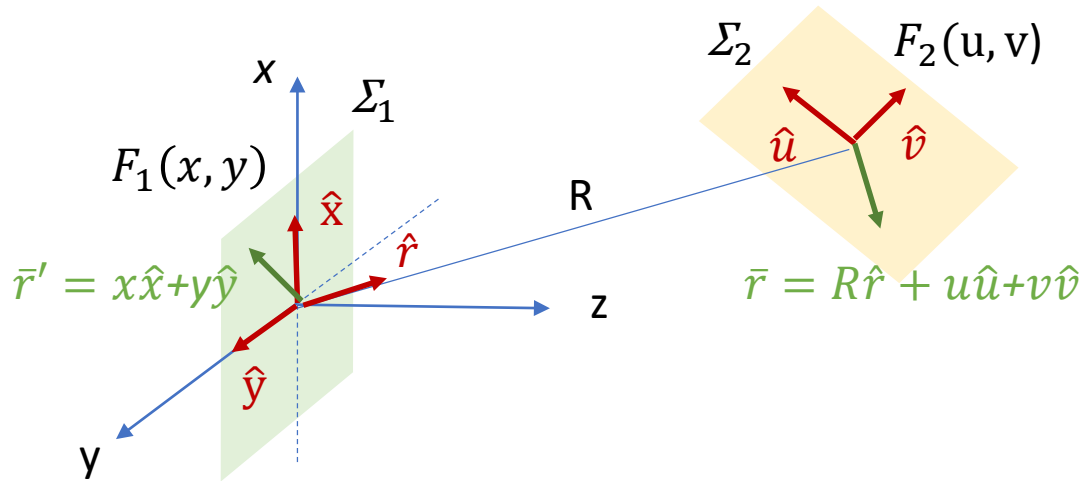


< LG mode >

- Beamforming for mobility
 - ✓ 5G beamforming for mobility
 - : Far-field region beamforming
 - : Polarized streams can be transmitted in each beam
 - ✓ 6G beamforming for mobility
 - : Fresnel region beamforming
 - : Multiple modes can be transmitted simultaneously in each beam
- Beamforming for misalignment
 - ✓ When Tx antenna and Rx antenna are not aligned, is it possible to obtain Fourier transform relationship between Tx signal and Rx signal?
 - ✓ Yes, beamforming can make Fourier transform relationship between Tx and Rx signals.

3 Beamforming for Mobility

- Relationship between Tx field and Rx field for misaligned Tx/Rx antenna arrays



$$a_{rx} = \hat{r} \cdot \hat{x}, a_{ry} = \hat{r} \cdot \hat{y}, a_{rz} = \hat{r} \cdot \hat{z}$$

$$a_{ru} = \hat{r} \cdot \hat{u}, a_{rv} = \hat{r} \cdot \hat{v}$$

$$a_{ux} = \hat{u} \cdot \hat{x}, a_{uy} = \hat{u} \cdot \hat{y}$$

$$a_{vx} = \hat{v} \cdot \hat{x}, a_{vy} = \hat{v} \cdot \hat{y}$$

$$\vec{r}' = x\hat{x} + y\hat{y}, (x, y) \in \Sigma_1, \vec{r} = R\hat{r} + u\hat{u} + v\hat{v}, (u, v) \in \Sigma_2$$

$$F_2(u, v) = \frac{j}{\lambda} \iint_{\Sigma_1} F_1(x, y) \frac{e^{-jk|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} a_{rz} dx dy$$

$$\approx \frac{je^{-jkz}}{\lambda R} e^{-jk(ua_{ru}+va_{rv})} e^{-\frac{jk}{2R}(u^2+v^2)}$$

$$\iint_{\Sigma_1} F_1(x, y) e^{jk(xa_{rx}+ya_{ry})} e^{-\frac{jk}{2R}(x^2+y^2)} e^{j\frac{k}{R}(uxa_{ux}+uya_{uy}+vxa_{vx}+vya_{vy})} a_{rz} dx dy$$

3 Beamforming for Mobility

- Tx/Rx beamforming for mobile users

$S_1(x, y)$: Tx signal $S_2(x, y)$: Rx signal

(1) Tx beamforming

$$F_1(x, y) = S_1(x, y) \underbrace{e^{-jk(xa_{rx}+ya_{ry})} e^{j\frac{k}{2R}(x^2+y^2)}}_{\text{Tx beamforming}}$$

(2) Rx beamforming

$$S_2(u, v) = F_2(u, v) \underbrace{e^{jk(ua_{ru}+va_{rv})} e^{-j\frac{k}{2R}(u^2+v^2)}}_{\text{Rx beamforming}}$$

(3) Relationship between Tx signal and Rx signal

$$S_2(u, v) = \frac{je^{-jkR}}{\lambda R} \iint_{\Sigma_1} S_1(x, y) e^{j\frac{k}{R}(uxa_{ux}+uya_{uy}+vxa_{vx}+vya_{vy})} a_{rz} dx dy$$

3 Beamforming for Mobility

- Coordinate transformation

$$\alpha = xa_{ux} + ya_{uy}$$

$$\beta = xa_{vx} + ya_{vy}$$

$$uxa_{ux} + uya_{uy} + vxa_{vx} + vya_{vy} = u\alpha + v\beta$$

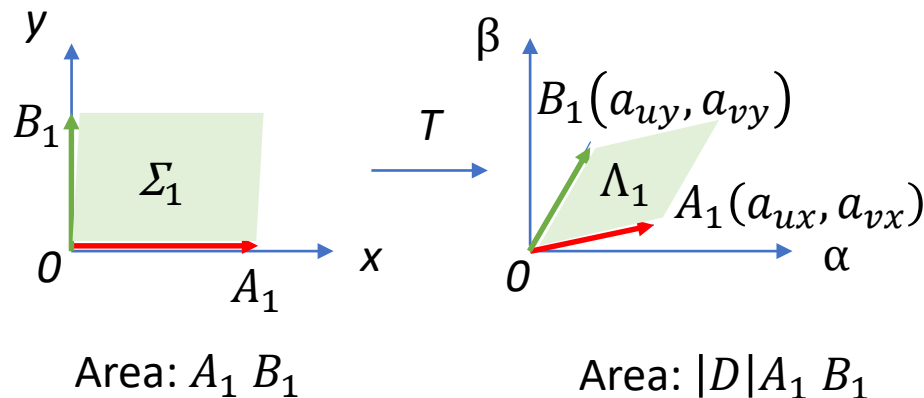
$$\begin{pmatrix} \alpha \\ \beta \end{pmatrix} = T \begin{pmatrix} x \\ y \end{pmatrix}, \quad T = \begin{pmatrix} a_{ux} & a_{uy} \\ a_{vx} & a_{vy} \end{pmatrix}$$

$$\begin{pmatrix} x \\ y \end{pmatrix} = T^{-1} \begin{pmatrix} \alpha \\ \beta \end{pmatrix}, \quad T^{-1} = \frac{1}{D} \begin{pmatrix} a_{vx} & -a_{uy} \\ -a_{vx} & a_{uy} \end{pmatrix}, \quad D = \begin{vmatrix} a_{ux} & a_{uy} \\ a_{vx} & a_{vy} \end{vmatrix} = a_{ux} a_{vy} - a_{uy} a_{vx}, \quad |D| \leq 1$$

$$Q_1(\alpha, \beta) = S_1(x, y) = S_1(T^{-1}(\alpha, \beta)) = S_1\left(\frac{1}{D}(a_{vy}\alpha - a_{uy}\beta), \frac{1}{D}(-a_{vx}\alpha + a_{ux}\beta)\right)$$

$$S_2(u, v) = \frac{je^{-jkR}}{\lambda R} \frac{a_{rz}}{|D|} \iint_{\Lambda_1} Q_1(\alpha, \beta) e^{j\frac{k}{R}(u\alpha + v\beta)} d\alpha d\beta$$

$$Q_1(\alpha, \beta) \xleftrightarrow{\text{2-dim. Fourier transform}} S_2(u, v)$$



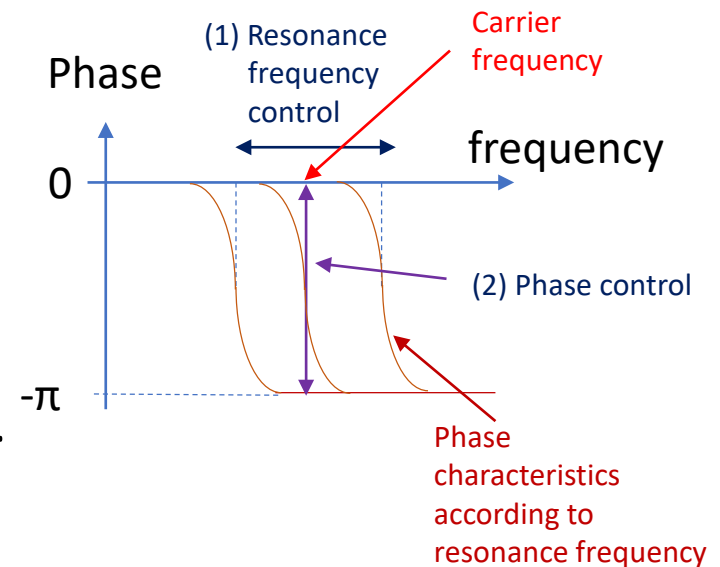
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Beamforming for Mobility

- Number of modes (rank)
 - ✓ Transmitter antenna array size: $A_1 * B_1$
 - ✓ Receiver antenna array size: $A_2 * B_2$
 - ✓ Number of modes: $\frac{A_1 B_1 A_2 B_2}{(\lambda R)^2} |D|$, $(D = a_{ux} a_{vy} - a_{uy} a_{vx})$
- Effect of non-parallel antenna arrays
 - ✓ Capacity may decrease if tx antenna array is not parallel to rx antenna array.
 - ✓ If Tx/Rx antenna arrays are parallel to each other,
then $|D| = 1$, and the number of modes and capacity are maximized.

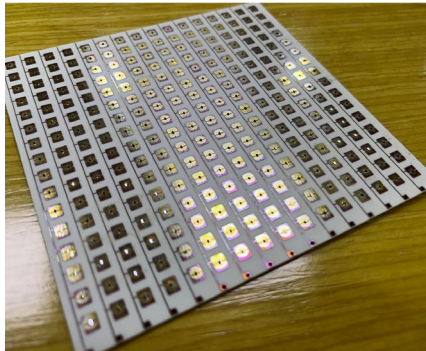
4 Reconfigurable Intelligent Surface

- Reconfigurable intelligent surface
 - ✓ Can overcome the shadowing problem in an NLoS environment.
 - ✓ Can be utilized for beamforming for mobile users.
 - ✓ Reflection type, transmittance type
- What should be considered about RIS?
 - ✓ RIS can be used in beamforming.
 - ✓ You should be able to change phases easily.
 - ✓ Controlling the amplitude is not that important.
- Principle of RIS
 - ✓ RIS can change the phase characteristics at the carrier frequency by controlling the resonance frequency.

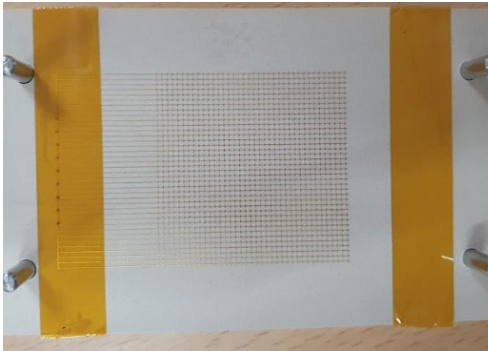


4 Reconfigurable Intelligent Surface

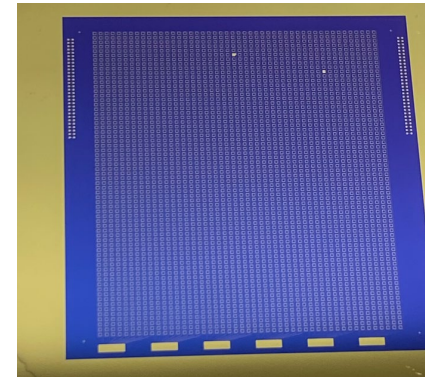
- How to control resonance frequency
 - ✓ Varactor diode: control capacitance (Upper-mid band, 24 GHz)
 - ✓ Graphene: control chemical potential (Sub-THz, 150 GHz, 300 GHz)
 - ✓ Liquid crystal: control permittivity (Sub-THz, 150 GHz, 300 GHz)



< Varactor based RIS >



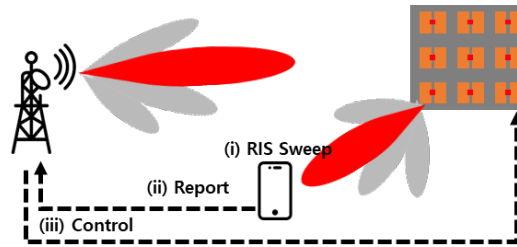
< LC based RIS >



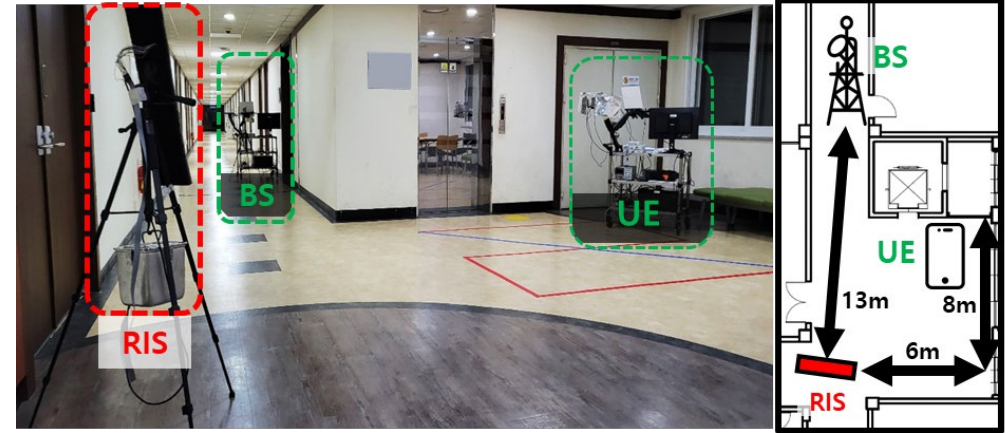
< Graphene based RIS >

4 Reconfigurable Intelligent Surface

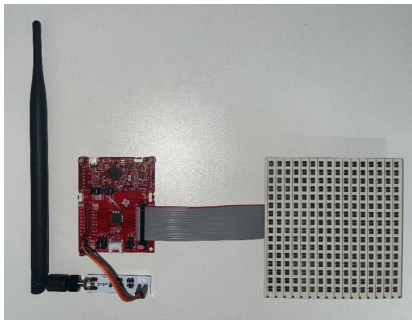
- RIS implementation in 6G upper-mid band (varactor diode)



< Concept of RIS beam control >



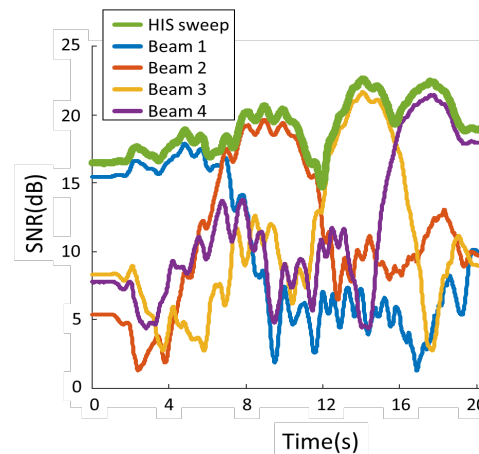
< Real-time RIS operation in indoor environment >



< RIS and RIS controller >



< Tx system (Rx system) >



< Measurement results >

- 1) Beam 1~4
- metasurface beamforming pattern 1~4
- 2) HIS sweep
- beamforming pattern giving the best SNR value

- Multi-mode transmission for 6G 1 Tbps capacity
 - ✓ Time and space resources should be utilized effectively.
 - ✓ In sub-THz band, the number of modes and capacity are determined by antenna array size and transmission distance.
 - ✓ Through transmitting multiple modes per each beam, we can achieve 1 Tbps capacity, which is one of the goals of 6G wireless communication.
- Beamforming for mobility
 - ✓ Beamforming is also required to serve mobile users.
 - ✓ There is a need to develop a low complexity method for multi-mode transmission and beamforming.
- RIS for 6G solution
 - ✓ RIS allows you to overcome the shadowing problem.
 - ✓ RIS can be effectively used in low complexity beamforming solutions.



Thank you